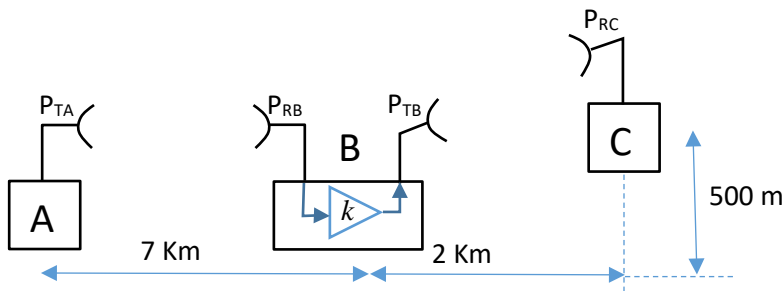


RF SYSTEMS

Written Test of February 11, 2019

Surname & Name
Identification Number
Signature

Exercise 1



The figure shows a terrestrial multi-link system operating at 6 GHz with bandwidth 150 MHz. All the antennas are identical and optimally directed, with efficiency $\eta=0.9$ and directivity function $f(\theta)=1/\cos^9(\theta)$.

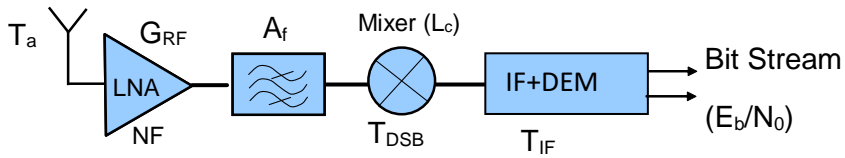
- 1) Evaluate the gain of the antennas (hint: $\int f^n(x) \cdot f'(x) dx = \frac{f^{n+1}(x)}{n+1}$)
- 2) Assume the power transmitted by A is $P_{TA}=30$ dBm. Assume also that the power transmitted by B is linearly related to the received power, i.e. $P_{TB}=k \cdot P_{RB}$. Evaluate the power P_{RB} and the gain k in dB in order that the power received in C is $P_{RC}=-80$ dBm
- 3) It is known that the receivers in B and C are characterized by $T_{eq}=150$ °K. Compute the overall SNR of the system defined as $SNR_{tot}=P_{RC}/KT_{sys}B$, with T_{sys} the overall system noise temperature (i.e. the overall noise computed at the input of receiver C).
- 4) If we increase the transmitted power P_{TA} by 3 dB and assume $T_{eq}=300$ °K, what is the new value of SNR_{tot} ?

Use $K=1.38 \cdot 10^{-23}$ (Boltzmann Constant)

Exercise 2

The scheme in the figure represents the RF front-end of the receiver C of the previous exercise.

The goal is to get $T_{eq}=150^{\circ}\text{K}$ with $P_{rec}=-80\text{ dBm}$.

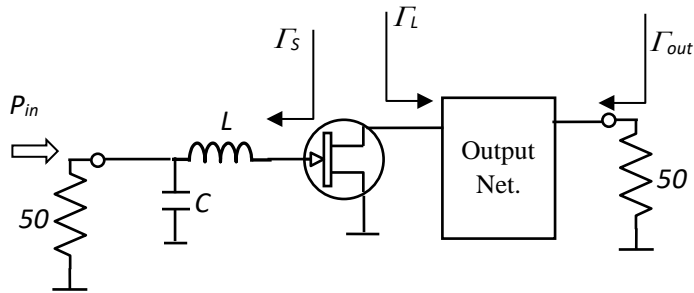


$NF = ?$
 $G_{RF} = 20\text{ dB}$
 $T_a = 100^{\circ}\text{K}$
 $A_f = 0.2\text{ dB}$
 $L_c = 4\text{ dB}$
 $T_{DSB} = 100^{\circ}\text{K}$
 $T_{IF} = 150^{\circ}\text{K}$

- Evaluate the Noise Figure NF_{RF} of LNA in order to meet the goal
- What is the maximum data rate R of the demodulated bit stream with $(E_b/N_0) = 10\text{ dB}$?
- If we assume $T_{eq} = 300^{\circ}\text{K}$ with $P_{rec} = -77\text{ dBm}$ and $NF = 1.5\text{ dB}$, what is the new requested value for G_{RF} ? Does the maximum data rate R change? (Justify the answer!)

Exercise 3

We want design a single stage amplifier at 1.8 GHz using the scheme in the following figure (the output network is assumed lossless):



The transistor has the following parameters:

$$S_{11}=0.52\angle 143.8^\circ \quad S_{21}=4.255\angle 28.7^\circ \quad S_{12}=0.081\angle 4.9^\circ \quad S_{22}=0.287\angle -87.7^\circ$$

$$\Gamma_{opt}=0.19\angle -170.4^\circ \quad F_{min}=0.85 \text{ dB} \quad R_n=0.06 \text{ (Noise parameters)}$$

$$P_{1dB}: 0 \text{ dBm}$$

The design goals are: $G_T=15 \text{ dB}$ and $NF=1 \text{ dB}$.

- 1) Evaluate Γ_S and Γ_L producing the requested values of G_T and NF . Note that the chosen value of Γ_S must be compatible with the assigned input network
- 2) Design the input network, i.e. determine L (nH) and C (pF)
- 3) The dynamic range (DR) is here defined as the ratio between the output power for a required back-off (BO) and the output power determining a specified input SNR. Evaluate DR for the designed amplifier with $BO=3 \text{ dB}$ and $SNR=15 \text{ dB}$ (assume the noise bandwidth $B=1 \text{ MHz}$)
- 4) What is the power at input for the maximum output power?
- 5) Assume that the amplifier is working at the nominal BO (3 dB) but the input signal is two-tone. Evaluate the carrier-to-intermodulation CI at output (assume ideal third order non-linearity). What is the output power in each tone?

Solution

Exercise 1

The antenna gain is obtained from the formula:

$$G = \eta 4\pi \left[\int_0^{2\pi} d\varphi \int_0^{\pi} f(\theta) \sin \theta d\theta \right]^{-1} = \frac{2\eta}{\int_0^{\pi} |\cos^9(\theta)| \sin(\theta) d\theta} = \frac{2\eta}{2 \left[-\frac{\cos^{10}(\theta)}{10} \right]_0^{\pi}} = 9 \text{ (9.54 dB)}$$

We have for the two links:

$$L_1 = 7000 \text{ m}, L_2 = (2000^2 + 500^2)^{0.5} = 2.0616 \text{ Km}, \lambda_0 = 3 \cdot 10^8 / 6 \cdot 10^9 = 0.05 \text{ m}$$

The Friis equation for the first link defines the received power P_{RB} :

$$P_{RB} = P_{TA} + 2G_A + 20 \log_{10} \left(\frac{\lambda_0}{4\pi L_1} \right) = -75.82 \text{ dBm}$$

For the second link it has:

$$P_{RC} = P_{TB} + 2G_A + 20 \log_{10} \left(\frac{\lambda_0}{4\pi L_2} \right) = -80 \text{ dBm} \rightarrow P_{TB} = 15.204 \text{ dBm}$$

The gain k is then given by: $k_{dB} = P_{TB} - P_{RB} = 91.03 \text{ dB}$

The overall noise temperature at input of C is given by:

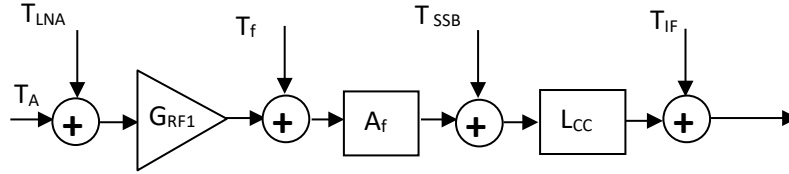
$$T_{sys} = T_C + T_B k G_A^2 \left(\frac{\lambda_0}{4\pi L_2} \right)^2 = 207.32 \text{ °K}$$

The system SNR is then given by:

$$SNR_{sys} = \frac{P_{RC}}{KT_{sys} B} = 23.3 \text{ (13.67 dB)}$$

Exercise 2

Equivalent noise scheme :



Where:

$$T_f = 293 \left(10^{\frac{A_f}{10}} - 1 \right) = 13.809 \text{ } ^\circ\text{K}, T_{SSB} = 2 \cdot T_{DSB} = 200 \text{ } ^\circ\text{K}$$

The equivalent noise temperature at the input of the receiver is then given by:

$$T_{eq} = T_A + T_{LNA} + \frac{T_f}{G_{RF}} + \frac{T_{SSB} A_f}{G_{RF}} + \frac{T_{IF} L_c A_f}{G_{RF}}$$

$$= T_{LNA} + 106.1778$$

Imposing $T_{eq} = 150$ we get:

$$T_{LNA} = 150 - 106.1778 = 43.8223 \text{ } ^\circ\text{K} \rightarrow NF = \left(1 + \frac{T_{LNA}}{T_0} \right)_{dB} = 0.605 \text{ dB}$$

The system SNR is given by:

$$SNR = \frac{P_{rec}}{K \cdot T_{eq} \cdot B} = \frac{E_B}{N_0} \left(\frac{R}{B} \right) = 15.08 \text{ dB}$$

$$\text{then: } (R/B)_{dB} = 15.08 - 10 = 5.08 \text{ dB} \rightarrow R = 3.22 \cdot B = 483.16 \text{ Mbit/sec}$$

With $T_{eq} = 300 \text{ } ^\circ\text{K}$ we have :

$$G_{RF} = \frac{T_f + T_{SSB} A_f + T_{IF} L_c A_f}{T_{eq} - T_A - T_{LNA}} = 7.807 \text{ (8.93 dB)}$$

The data rate R does not change (both P_{rec} and T_{eq} double so SNR remains constant)

Exercise 3

The transistor is unconditionally stable with $G_{max}=16.8$ dB and $NF_{min}=0.85$.

Draw the circles $NF=1$ dB and $G_{AV}=15$ dB on the S.C. Γ_S is chosen between the two intersections of these circles. We must however verify that the selected value is compatible with the input network. This happens for $\Gamma_S = 0.48 \angle -154.9^\circ$. The value of Γ_L is the one which produces conjugate matching at output (in this way $G_T=G_{AV}$). With the S.C. we find $\Gamma_L = \text{Optimum Gamma Load} = 0.48 \angle 103.67^\circ$.

The input network can be designed either with analytic formulas or with the S.C. It is found:
 $X=50 \cdot 0.286=14.3 \Omega$, $B=0.02 \cdot 1.33=0.0266$. From which: $L=X/\omega_0=1.2644$ nH, $C=B/\omega_0=2.352$ pF

The maximum output power for the given BO and P1dB is $P_{max}=-3$ dBm. The minimum power is defined as:

$P_{min}=P_{in}-G_T$, with $P_{in}=P_N-SNR$, $P_N=KT_{eq}B$, $T_{eq}=T_0(10^{(NF/10)}-1)=75.865$ °K.

Replacing we get:

$P_N=1.047 \cdot 10^{-15}$ W (-119.8 dBm), $P_{in}=-119.8+15=-104.8$ dBm, $P_{min}=-104.8+15=-89.8$ dBm

Then:

$DR=P_{max}-P_{min}=-3+89.8=86.8$ dB

In case of two-tone signal the output average power remain -3 dBm and the power per tone is -6 dBm. The CI is obtained with the following formula ($\Delta_p=10.63$):

$CI = 2(BO + \Delta_p + 3) = 33.26$ dB