RF SYSTEMS

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Exercise 1 (8+2)

Consider the antenna used in a base station for LTE service, placed at H=30 m above ground. The operating (downlink) frequency is f_0 =2.12 GHz.



The directivity function is expresses as:

$f\left(\varphi,\theta\right) = f_{\varphi}\left(\varphi\right) \cdot f$	$f_{ heta}(heta)$	φ : azimut	h, θ	elevation:
$f_{\varphi}(\varphi) = \cos(\varphi)$	$-\pi/2$	$\leq \varphi \leq \pi/2$,	0	elsewhere
$f_{\theta}(\theta) = \sin(5\theta)$	$2\pi/5$	$\delta \le \theta \le 3\pi/5,$	0 0	elsewhere

Note that angles θ and φ refer to the pointing direction ($\theta = 90^\circ$, $\varphi = 0^\circ$). This direction is tilted by θ_{tilt} with respect to the direction parallel to ground in order to reduce the interference among adjacent cells.

- 1) Specify the angles θ_{max} and φ_{max} where $f(\theta, \varphi)$ is maximum
- 2) Compute the antenna gain (efficiency factor η =0.85).

Hint: $\int \sin(x)\sin(5x) = \frac{\sin(4x)}{8} - \frac{\sin(6x)}{12}$

- 3) x_s represents the distance from the base station at ground level determined by antenna tilting (see the figure). Imposing $x_s=1.5$ Km, determine the value of θ_{tilt}
- 4) The transmitter power level is 30 dBm. Evaluate the power density (W/m²) at x_S (with $\phi = \phi_{max}$).

(Optional)

- 5) Evaluate the angles θ_{3dB} and ϕ_{3dB} where f_{ϕ} and f_{θ} are reduced by -3dB with respect to the optimum pointing (Hint: there are two values for each angle)
- 6) Evaluate also the power density at x_{3dB} (the point at ground level determined by θ_{3dB}). Assume $\phi = \phi_{3dB}$

Exercise 2 (8)

The scheme in the figure represents the RF front-end of a phone receiver using the BS of the previous exercise.



- a) Draw the scheme of the receiver referred to the equivalent noise sources. Evaluate the equivalent noise temperature (T_{eq}) at the receiver input. Compute the noise power at the receiver input (assume B=20 MHz). Use K=1.38 \cdot 10^{-23} (Boltzmann Constant)
- b) Write the expression of the received power (at the antenna output) as function of the distance R from the BS
- c) It is requested the data rate R=300 Mbit/sec, with $(E_b/N_0) = 10$ dB. Evaluate the corresponding SNR of the receiver.
- d) Evaluate the maximum distance *R* from the BS at which the data rate R is still guaranteed.

Exercise 3 (8)

The following scheme represents an amplifier operating at 12 GHz.

a) Determine the values Γ_{s} and Γ_{L} for the maximum transducer gain

b) Evaluate the unknown parameters (Z_{c1} , Z_{c2} , βI_1) of the output matching network



Scattering parameters of the transistor:

 $S_{11} = 0.66 \angle 146^{\circ} \quad S_{21} = 2.39 \angle 45^{\circ} \quad S_{12} = 0.088 \angle 69^{\circ} \qquad S_{22} = 0.3 \angle -48^{\circ}$

Exercise 4(9)

The following scheme refers to an oscillator working at 432 MHz. The S parameters of the transistor are also reported on the figure.



- 1) Select a suitable value for L_S (use the mapping circles of Γ_s for obtaining $|\Gamma_{out}|$ =1.5)
- 2) Evaluate the parameters of the output network (jX, jB) to ensure the start of oscillation and the transfer of the output power to the external load (50 Ω).
- 3) Find the values of the lumped elements parameters implementing X and B (select an inductor or a capacitor depending on the sign of X and B)

Solution

Exercise 1

1) $\theta_{max}=90^{\circ}$, $\phi_{max}=0^{\circ}$

2) The antenna gain is obtained from the formula:

$$G = \eta 4\pi \left[\int_{0}^{2\pi} d\varphi \int_{0}^{\pi} f(\theta) \sin \theta d\theta \right]^{-1} = \frac{4\pi\eta}{\int_{-\pi/2}^{\pi/2} \cos(\varphi) d\varphi} \int_{2\pi/5}^{3\pi/5} \sin(5\theta) \sin(\theta) d\theta = \frac{4\pi\eta}{0.7925} = 13.4773$$
3) $x_s = H \cdot \tan(90 - \theta_{tilt}) \Rightarrow \theta_{tilt} = 90 - \tan^{-1}\left(\frac{x_s}{H}\right) = 1.146^{\circ}$

$$R_s = \frac{x_s}{\cos(\theta_{tilt})} = 1500.3 \text{ m}$$
4) $S_R = \frac{P_r G}{4\pi R_s^2} = \frac{1 \cdot 13.4773}{4\pi \cdot 1500.3^2} = 4.765 \cdot 10^{-7} \text{ W/m}^2$

$$f_{\varphi}(\varphi_{3dB}) = 0.5 \Rightarrow \varphi_{3dB} = \pm \pi/3$$
5) $f_{\theta}(\theta_{3dB}) = 0.5 \Rightarrow \theta_{3dB} = \frac{1}{5}\left(\frac{\pi}{6} + 2\pi\right) = \frac{13}{30}\pi$

$$= \frac{1}{5}\left(\frac{5\pi}{6} + 2\pi\right) = \frac{17}{30}\pi$$

The above angles are those referred to the antenna pointing. Taking into account the tilting, the angle θ'_{3dB} with respect the absolute coordinate is given by $\theta'_{3dB} = \theta_{3dB} + \theta_{tilt}$

6) The distance R_{3dB} is given by: $R_{3dB} = H/\cos(\theta'_{3dB} - 90) = 131.9 \text{ m}$. Then:

$$S_{R_{3dB}} = \frac{P_T G}{4\pi R_{3dB}^2} f(\theta_{3dB}) f(\varphi_{3dB}) = 1.548 \cdot 10^{-5} \text{ W/m}^2$$

Exercise 2

a) Equivalent noise scheme :



Where:

$$T_f = 293 \left(10^{\frac{A_f}{10}} - 1 \right) = 171.4 \text{ }^{\circ}\text{K} \text{, } T_{LNA} = 293 \left(10^{\frac{NF}{10}} - 1 \right) = 1175.5 \text{ }^{\circ}\text{K}$$

The equivalent noise temperature at the input of the receiver is then given by:

$$T_{eq} = T_A + T_{LNA} + \frac{T_f}{G_{RF}} + \frac{T_{SSB}A_f}{G_{RF}} + \frac{T_{IF}A_cA_f}{G_{RF}} = 1887.3 \text{ °K}$$
$$P_N = KT_{eq}B = 5.21 \cdot 10^{-13} \text{ W}$$

b) Friis equation:

$$P_r = P_T \cdot \left(\frac{\lambda}{4\pi R}\right)^2 \cdot G_T \cdot G_R = 0.002152/R^2$$

c) Evaluation of SNR:

$$SNR = \frac{P_r}{P_N} = \left(\frac{E_b}{N_0}\right) \left(\frac{R}{B}\right) = 150$$

d) Evaluation of the distance R: $P_r = SNR \cdot P_N = 7.815 \cdot 10^{-11} = 0.002152/R^2$ $R = \sqrt{\frac{0.002152}{7.815 \cdot 10^{-11}}} = 5.248 \text{ Km}$

Exercise 3

The transistor is unconditionally stable with Gmax=12.1 dB. The optimum gamma are: $\Gamma_S=0.84 \angle -147.5^\circ$, $\Gamma_L=0.69 \angle 40.8^\circ$. $\beta I_1=139.2/2=69.6^\circ$, $Z_{c1}=50\Omega$.

$$Z_{in} = 50 \cdot 0.18 = 9.05$$
$$Z_{c2} = \sqrt{Z_{in} \cdot 50} = 21.27 \ \Omega$$

Exercise 4

The mapping circle of Γ s is drawn on the S.C with $|\Gamma$ out|=1.5. The two intersections with the outer circle are: Γ 1=1 \angle 120.36, Γ 2=1 \angle 81.54.

We assign $\Gamma s = \Gamma 2$, from which $Zs = j1.16 \cdot 50 = j58 \Omega$. The inductance Ls is then derived as:

$$L_s = \frac{X_s}{2\pi f_0} = 21.37 \text{ nH}$$

At the transistor out we get $\frac{Z_{out}}{50} = -0.726 - j1.5$.

Imposing:
$$Z_L = \frac{-\text{Re}(Z_{out})}{3} - \text{Im}(Z_{out})$$
 we get $Z_L = 50 \cdot (0.24 + j1.5)$.

The design of the matching network can be carried out either with the S.C. or by means of the formulas. The following results are obtained:

X=1.073[·]50=53.64Ω, B=-1.776/50=-0.03552 S.

The lumped components value is then computed as follows:

$$L_{X} = \frac{|X|}{2\pi f_{0}} = 19.76 \text{ nH}$$
$$L_{B} = \frac{1}{|B|2\pi f_{0}} = 10.37 \text{ nH}$$